

PSO Training Neural Network MPPT with CUK Converter Topology for Stand-Alone PV Systems Under Varying Load and Climatic Conditions

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(Received: 22.01.2024, Accepted: 01.03.2024, Online Publication: 29.03.2024)

Keywords

Maximum power point tracking, Particle swarm optimization based neural network, CUK converter, Boost converter

Abstract: Temperature and irradiance levels are two examples of environmental variables that affect the power value produced by photovoltaic panels. Therefore, in order to transfer the maximum power value from the PV panel to the load under varying climatic conditions, maximum power point tracking (MPPT) algorithms and DC-DC converter topologies are used. In this study, the performances of boost converter and CUK converter circuit topologies are investigated under variable irradiance and variable load conditions by using a neural network-based MPPT algorithm learning particle swarm optimization (PSO). As the first scenario, it is analyzed assuming that the temperature and irradiance values coming to the panel are constant. As the second scenario, the performance evaluation of the converter topologies according to the current, voltage and power parameters is made for the variable load situation. As the last scenario, the difference in the irradiance value coming to the panel depending on the sun's condition during the day has been examined. Canadian Solar CS6P-250P PV panel is used in the study. 50 kHz is selected as the switching frequency. According to the results obtained, it has been observed that the CUK converter circuit topology reaches the maximum power point faster than the boost converter circuit topology both in dynamic environmental conditions and load change, and the oscillation at this point is less. It is aimed to increase the performance of this method, which uses boost converter topology and MPPT in the literature, by applying CUK converter topology.

Değişken Yük ve İklim Koşulları Altında Müstakil Çalışan PV Sistemleri için CUK Dönüştürücü Topolojili PSO Eğitilmiş Sinir Ağı Tabanlı MPPT

Anahtar

Kelimeler

Maksimum güç noktası takibi, Parçacık sürü optimizasyon tabanlı sinir ağı, CUK dönüştürücü, Yükselten tip dönüştürücü

Öz: PV panellerden elde edilen güç değeri sıcaklık ve ışınım değerleri gibi çevresel faktörlere bağlı olarak değişmektedir. Bu nedenle değişen iklim koşullarında PV panelden maksimum güç değerinin yüke aktarılması için maksimum güç noktası izleme (MPPT) algoritmaları ve DC-DC dönüştürücü topolojileri kullanılmaktadır. Bu çalışmada, parçacık sürü optimizasyonu (PSO) öğrenen sinir ağı tabanlı MPPT algoritması kullanılarak, yükselten tip dönüştürücü ve CUK dönüştürücü devre topolojilerinin performansları değişken ışınım ve değişken yük koşulları altında incelenmiştir. İlk senaryo olarak panele gelen sıcaklık ve ışınım değerlerinin sabit olduğu varsayılarak analiz edilmiştir. İkinci senaryo olarak değişken yük durumu için akım, gerilim ve güç parametrelerine göre dönüştürücü topolojilerinin performans değerlendirmesi yapılmıştır. Son senaryo olarak ise gün içerisinde güneşin durumuna bağlı olarak panele gelen ışınım değerindeki farklılık incelenmiştir. Çalışmada Canadian Solar CS6P-250P PV panel kullanılmıştır. Anahtarlama frekansı olarak 50 kHz seçilmiştir. Elde edilen sonuçlara göre CUK dönüştürücü devre topolojisinin hem dinamik çevre koşullarında hem de yük değişiminde yükselten tip dönüştürücü devre topolojisine göre maksimum güç noktasına daha hızlı ulaştığı ve bu noktadaki salınımın daha az olduğu görülmüştür. Literatürde yükselten tip dönüştürücü topolojisi ve MPPT kullanan bu yöntemin performansının CUK dönüştürücü topolojisi uygulanarak artırılması hedeflenmiştir.

1. INTRODUCTION

Due to factors such as technological developments, increasing population and developing industry, the need for electrical energy is increasing over time. Fossil energy fuels, which are used to meet the energy needs, that are of primary importance in the development of countries, are gradually depleted. The use of fossil energy fuels is decreasing day by day due to the harmful gases they emit to the environment. Renewable energy sources have been used in order to meet the increasing energy demand in recent years. The most important advantages of these resources are that they are clean, constantly renewable, reduce environmental pollution and have low operating costs. The most important renewable energy sources used today are hydroelectric energy, wind energy, solar energy, geothermal energy and wave energy. Solar energy systems are preferred because they have no fuel costs, contain no moving parts, are resistant to climate changes and directly convert solar energy into electrical energy. Photovoltaic (PV) panels are the most important part of solar energy systems. PV panels are obtained from the combination of PV cells. PV panels are connected in series or parallel to create an energy system at the desired power level. Another important factor affecting the PV panel efficiency is the amount of solar irradiance reaching to the panel. During the day, the amount of irradiance reaching to the panel is reduced in cases such as cloudy weather and tree shadows. As a result, the amount of power produced decreases and the panel efficiency decreases. For these reasons, MPPT algorithms are used to increase panel efficiency. MPPT algorithms basically try to provide maximum energy to the load by optimizing the duty ratio of the DC-DC converters that provide the connection between the PV panel and the load. MPPT control algorithms are divided into two groups as analog and digital control. MPPT is provided with the help of conventional controllers by comparing the reference voltage and output voltage produced PV panel in analog control. In the digital control method, the control signal is produced directly by the pulse width modulation (PWM) generator.

In both control methods, the aim is to transfer the power obtained from the PV panel to the load with the highest possible efficiency through optimization algorithms. Optimization algorithms are classified as classical and metaheuristic. Classical optimization algorithms are frequently preferred in real-time applications due to their advantages such as simple structure and easy applicability. The most used optimization algorithms for MPPT applications are perturb and observe algorithm (P&O) [1-8], incremental conductance algorithm (INC) [9]. MPPT is preferred as analog control method in fuzzy logic-based algorithms (FLC) [3,10-11], neural network-based (NN) [12-16] and machine learning based [17] algorithms. Under partial shading conditions, bypass diodes cause multiple local maximum points in the P-V characteristic. In this case, classical optimization algorithms cannot reach the global maximum because they follow the local maximum. Metaheuristic

optimization algorithms are used to increase PV panel efficiency. Major metaheuristic optimization algorithms; African vulture optimization algorithm (AVOA) [18], crow search algorithm (CSA) [19], butterfly optimization algorithm (BOA) [20], whale optimization algorithm (WOA) [21], most valuable player algorithm (MVPA) [22], squirrel search algorithm (SSA) [23], shuffled frog-leaping algorithm (SFLA) [24,25], BAT search [26], Harris hawk optimization (HHO) [27], search and rescue algorithm (SRA) [28] and particle swarm optimization (PSO) [7,29-33]. There are comparisons of some metaheuristic optimization algorithms based on performance parameters such as complexity, efficiency, and convergence time [34,35]. Al-Majidi et al designed a feedforward artificial neural network (ANN) for MPPT. In order to increase the accuracy of the model designed for MPPT, the PSO algorithm is used for the initial weights and selection of the best topology. While training, irradiance values and temperature values are determined as input variables, and power value is determined as output variable. In order to prove the accuracy of the designed model, a real-time data collection system is created. With this system, 48500 data are collected in a year on sunny and cloudy days. The suggested approach is compared with P&O, FLC, and traditional ANN algorithms. The outcomes gained show that the proposed strategy is successful. A boost converter topology is recommended to raise the input voltage produced by the PV panel [36]. In this study, it is aimed to increase the performance of the proposed method by applying it to amplifying type and CUK converter circuit topologies. This paper is structured as follows: PV cell model, boost converter, CUK converter, PSO, NN and PSO training NN are explained in Sect. 2. The simulation model created to compare the performance of the boost converter and CUK converter circuit topologies and the results are presented comparatively in Sect. 3. Finally, conclusions are given in Sect. 4.

2. MATERIAL AND METHOD

2.1. PV Cell Model

PV cells are obtained from the combination of p-n semiconductor elements. PV cells are semiconductor materials that generate current depending on the irradiance value. The most frequently used single-diode PV cell model in the literature for modeling PV cells is shown in Figure 1 [37]. I_{pv} is the current produced by light photons, I_d is the diode current, G is the irradiance value from the sun, T is the temperature value, V_p is the voltage value obtained from the PV panel. In materials with a thin film structure made up of extremely thin layers, R_p represents the total of the resistances formed between the layers and surrounding the cell, and R_s represents the sum of the resistances of the semiconductor material forming the cell and the contact resistances formed at the junction points of the cells.

Nomenclature	Abbreviation		
PV	Photovoltaic	BOA	Butterfly optimization algorithm
MPPT	Maximum power point tracking	WOA	Whale optimization algorithm
PWM	Pulse width modulation	MVPA	Most valuable player algorithm
P&O	Perturb and observe	SSA	Squirrel search algorithm
FL	Fuzzy logic	SFLA	Shuffled frog-leaping algorithm
P&O	Perturb and observe	HHO	Harris hawk optimization
INC	Incremental Conductance	SRA	Search and rescue algorithm
NN	Neural networks	PSO	Particle swarm optimization
CSA	Cuckoo search algorithm	ANN	Artificial neural network
BLDC	Brushless direct current	ML	Machine Learning
AVOA	African vulture optimization algorithm		

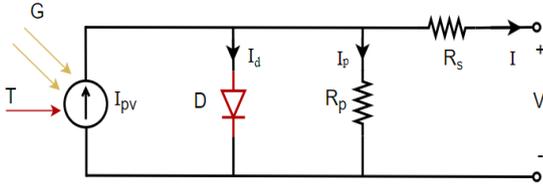


Figure 1. Single diode PV cell model [16]

The equations for modeling PV panels are given in Equations (1-5).

$$I = I_{pv} - I_d - I_p \tag{1}$$

$$I_d = I_0 \left(e^{\frac{qV_d}{kT}} - 1 \right) \tag{2}$$

$$V_d = V + IR_s \tag{3}$$

$$I_d = I_0 \left(e^{\frac{q(V+IR_s)}{nkT}} \right) \tag{4}$$

$$I = I_{pv} - I_0 \left(e^{\frac{q(V+IR_s)}{nkT}} \right) - \frac{V + IR_s}{R_p} \tag{5}$$

Where n represents the diode ideality constant, T represent for the cell temperature (K), k is the Boltzman constant (J/K), q is the electron charge (C) and I_0 represents reserve saturation current (A). Canadian Solar CS6P-250P was used as PV panel in the study. The features of this panel are given in Table 1.

Table 1. Canadian Solar CS6P-250P

Parameters	Value
Maximum Power	249.83 W
Open Circuit Voltage	37.2 V
Vmp	30.1 V
Isc	8.87 A
Imp	8.3 A
Temperature coefficient of Isc (%/deg.C)	0.063698
Temperature coefficient of Voc (%/deg.C)	-0.3654
Ncell	60

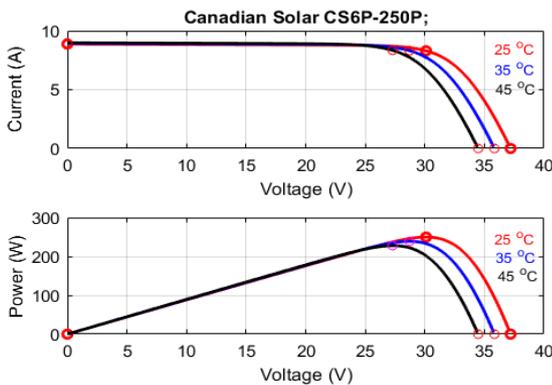


Figure 2. PV system curves at fixed 1000 W/m² irradiance value

The two most significant factors influencing PV panel output power are temperature and irradiance levels. Figure 2 shows the voltage dependent variation of the output power and current of the Canadian Solar CS6P-250P PV panel according to the constant irradiance and variable temperature values. It is seen that as the temperature value increases, the voltage value decreases and the power value increases in Figure 2.

Figure 3 illustrates the variation of the output power and current depending on the voltage, according to the constant temperature and variable irradiance values. As can be seen from Figure 3, the current value decreases as the irradiance value decreases. In addition, since the power value depends on the voltage and current values, it is seen that there is a decrease in this value.

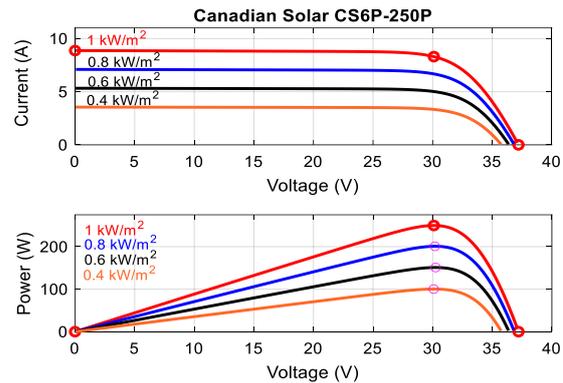


Figure 3. PV system curves at fixed 25°C temperature value

2.2. Boost Converter Model

Step up circuit topologies that transfer the unadjusted DC voltage value applied to the input to the output at higher levels and adjustable. It is also known as step up converter. It operates in two different states depending on the open or close state of the switching element. In the case of switching is ON-mode, the diode is polarized in the opposite direction and shows open circuit property. In the case of switching is OFF-mode, the diode is forward-biased and shows short-circuit characteristics. The boost converter circuit topology is shown in Figure 4 [16].

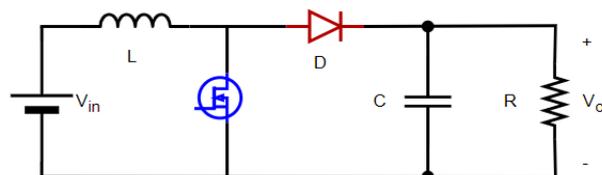


Figure 4. Boost converter circuit topology

The boost converter parameters are determined using the following equations. Average output current;

$$I_0 = \frac{P}{V_0} \quad (6)$$

$$\Delta I = (0.05) \frac{I_0 V_0}{V_{in}} \quad (7)$$

$$\Delta V = V_0(0.01) \quad (8)$$

Inductance;

$$L = \frac{V_{in} (V_0 - V_{in})}{f \Delta I V_0} \quad (9)$$

Capacitance;

$$C = \frac{V_{in} (V_0 - V_{in})}{f \Delta I V_0} \quad (10)$$

The circuit parameters of the boost converter used in simulation studies are given in Table 2.

Table 2. The parameters of Boost converter topology

Parameters	Value
L	1 mH
C _{in} and C ₂	100 μF- 42.5 μF
R _L	30 Ω
Switching Frequency (f)	50 kHz

2.3. CUK Converter Model

CUK circuit topologies that transfer the DC voltage value applied to the input to the output at higher or lower levels. Today, CUK converters are used in brushless direct current (BLDC) motor driver circuits, renewable energy systems and pulse width modulation (PWM) based PV systems. The most important difference of CUK converters from other converters is the use of capacitors for energy transfer. The CUK converter circuit topology is shown in Figure 5.

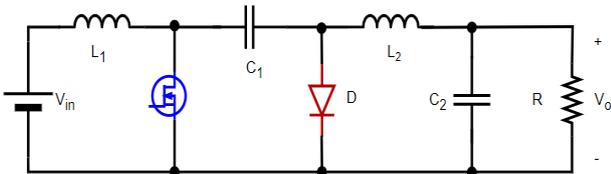


Figure 5. CUK converter circuit topology

The following equations are used in the CUK converter design given in Figure 5;

Duty ratio;

$$D = - \frac{V_0}{V_{in} - V_0} \quad (11)$$

Average inductance currents;

$$i_{L1} = \frac{P_s}{V_{in}} \quad (12)$$

$$i_{L2} = \frac{P_s}{-V_0} \quad (13)$$

Rate of change inductance currents;

$$\Delta i_{L1} = i_{L1} (\%10) \quad (14)$$

$$\Delta i_{L2} = i_{L2} (\%10) \quad (15)$$

Inductances;

$$L_1 = \frac{V_{in} D}{f \Delta i_{L1}} \quad (16)$$

$$L_2 = \frac{V_{in} D}{f \Delta i_{L2}} \quad (17)$$

Rate of change capacitance voltage;

$$\Delta V_{C1} = (V_{in} - V_0) 0.01 \quad (18)$$

Load resistance;

$$R = \frac{V_0^2}{P} \quad (19)$$

Capacitances;

$$C_1 = \frac{V_0 D}{R f \Delta V_{C1}} \quad (20)$$

$$C_2 = \frac{1 - D}{0.01 f^2 8 L_2} \quad (21)$$

where we have; V_{in} input voltage (V), V_0 output voltage (V), i_L average inductance current (A), D duty ratio, Δi_L rate of change inductor current, L inductance (H), C capacitance (F), ΔV_c rate of change capacitor voltage, R load resistance (Ω), f switching frequency (Hz), r ripple, r_{c1} ripple at C_1 .

Circuit parameters of the CUK converter used in simulation studies are given in Table 3.

Table 3. The parameters of CUK converter

Component	Parameters
L ₁ and L ₂	5.74 μH- 1.59 μH
C ₁ and C ₂	39.5 μF- 8.36 μF
R _L	30 Ω
Switching Frequency	50 kHz

2.4. Particle Swarm Optimization

The PSO algorithm was first developed in 1995 by R.C. It was introduced by Eberhart and J. Kennedy. The basis of the algorithm is based on the behavior of flocks of birds. Swarms are made up of particles. Each particle adjusts its position to its best position, taking advantage of its previous experience. Afterwards, all particles update their position according to the best particle of the swarm. This process continues until the goal is reached. In Figure 6, the flow chart of the PSO algorithm is given. The particle velocity and position are shown in the equations below. The velocity and position information of the particles are calculated according to Equation 22 and Equation 23 [31].

$$v_{k+1}^i = w v_k^i + c_1 r \text{and} \frac{p^i - x_k^i}{\Delta t} + c_2 r \text{and} \frac{p_k^g - x_k^i}{\Delta t} \quad (22)$$

$$x_{k+1}^i = x_k^i + v_{k+1}^i \quad (23)$$

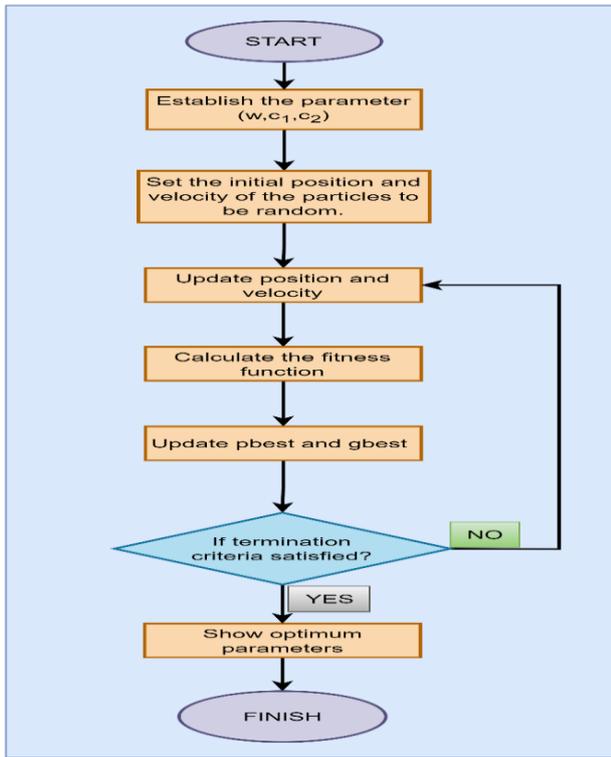


Figure 6. PSO flowchart

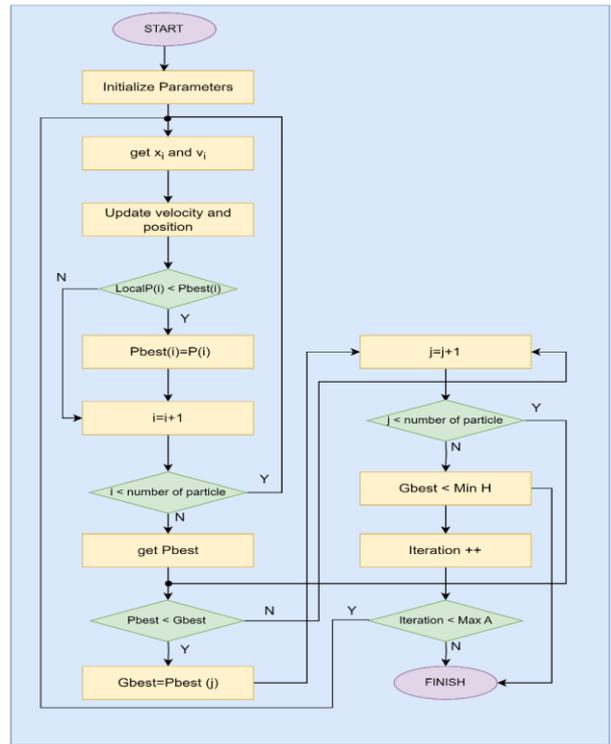


Figure 8. PSO training NN Algorithm [38]

2.5. Neural-Network Model

The first artificial neural network model was created in 1943 by W. McCulloch and W. Pitts. In general, neural networks have usage areas such as system modeling, handwriting recognition, automatic vehicle control, voice recognition, fingerprint recognition and meteorological interpretation. An artificial neuron is shown in Figure 7.

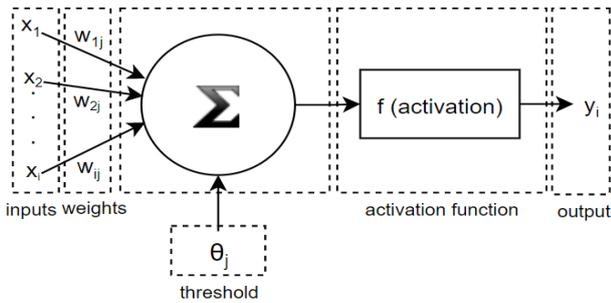


Figure 7. Artificial neuron

2.6. PSO training Neural-Network Model

PSO algorithm is used in training neural networks in order to obtain faster training process and convergence time. The algorithm is started by choosing a random location for each particle. The velocity and position information of each particle is updated using Equation 22 and Equation 23 by creating loops as many as the number of particles. Then the *Gbest* value is updated by comparing it with the *Pbest* values in each iteration. Finally, the algorithm is terminated by looking at the stopping criteria. The flowchart of NN training with PSO is shown in Figure 8.

3. RESULTS AND DISCUSSION

The simulation model shown in Figure 9 was created to compare the performances of the boost converter and the CUK converter. While creating the simulation model, Canadian Solar CS6P-250P was chosen as the PV panel. Boost converter and CUK converter circuit topologies are utilized as DC-DC converter architectures to raise the voltage at the panel output. 30Ω resistive as load, PSO trained neural network maximum power point tracking (NN MPPT) algorithm is used to optimize the duty ratios of converter topologies. The irradiance values were determined as 1000 W/m², 800 W/m², 600 W/m², 400 W/m² in the first case, constant 700 W/m² in the second case and a *trapezoidal function* in the last case. Irradiance values for different scenarios are shown in Figure 10.

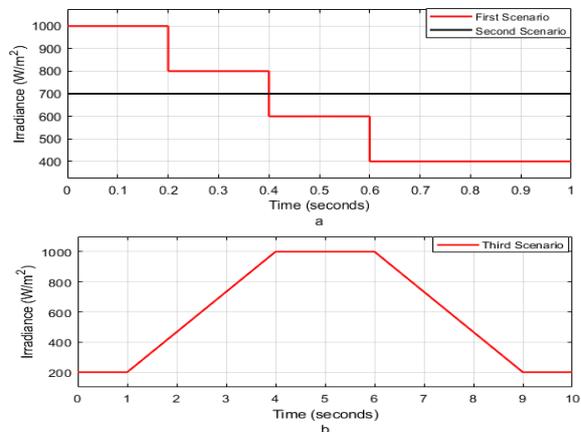


Figure 10. Irradiance values for a) first and second scenarios, b) third scenario

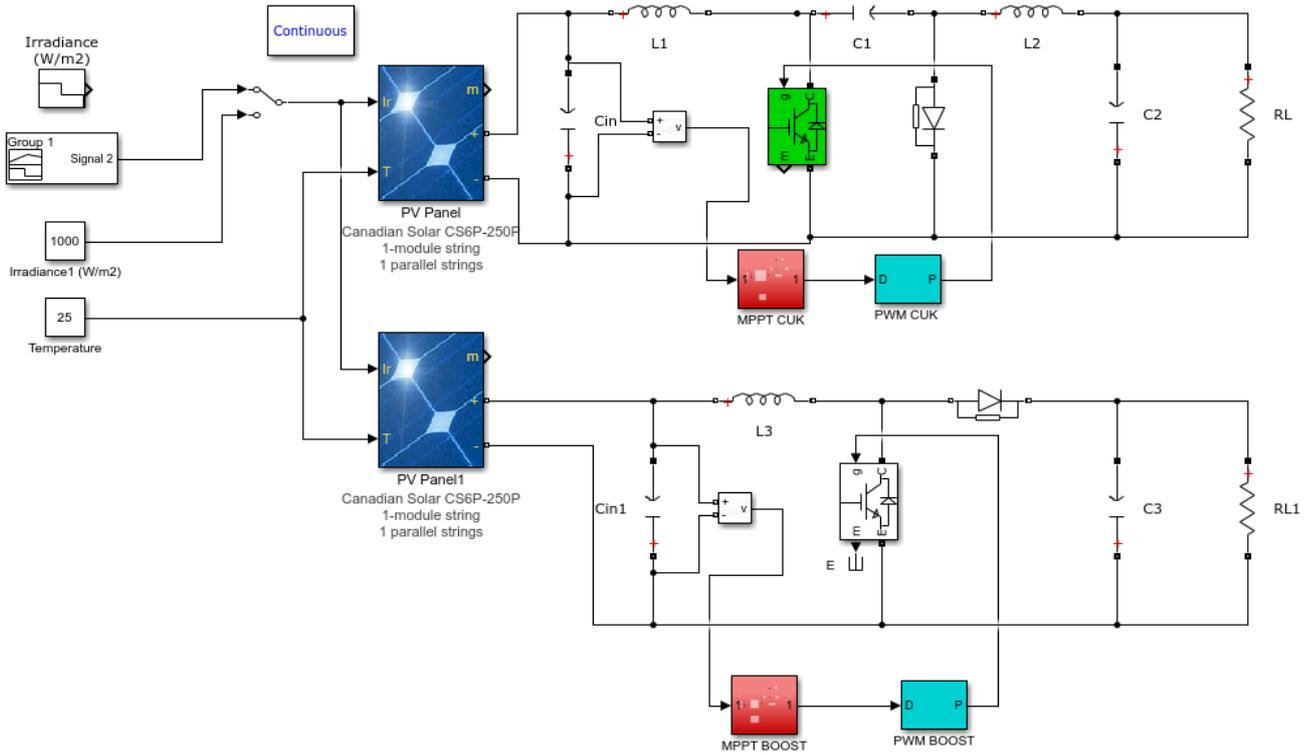


Figure 9. Simulation Model

First Scenario: In the first scenario, the irradiance value was initially selected as 1000 W/m^2 and the converter performances were examined by reducing 200 W/m^2 in 0.2 s intervals. The irradiance value was determined as 800 W/m^2 in the $0.2-0.4 \text{ s}$ time interval, 600 W/m^2 in the $0.4-0.6 \text{ s}$ time interval and 400 W/m^2 in the $0.6-0.9 \text{ s}$ time interval. Irradiance values are chosen as high, medium and low irradiance levels and the changes in current, voltage and power at these values are investigated. The temperature was determined as 25°C fixed values. Figure 11 shows the time-dependent variation of the load currents of the boost converter and CUK converter circuit topologies.

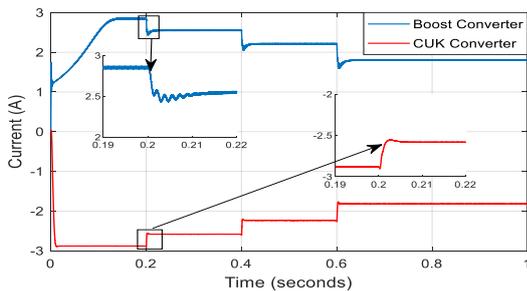


Figure 11. Boost converter and CUK converter output currents

In Table 4, the current values of both converters are given for the scenario determined above.

Table 4. The performance of Boost and CUK converter topologies

Time	Irradiance Value	Boost Converter	CUK Converter
0-0.2s	1000 W/m^2	2.856A	-2.88A
0.2-0.4s	800 W/m^2	2.558A	-2.58A
0.4-0.6s	600 W/m^2	2.21A	-2.24A
0.6-1s	400 W/m^2	1.81A	-1.83A

In Figure 12, the change in the voltage values of the boost converter topology and the CUK converter topology due to the decrease in the irradiance value at certain time intervals is shown.

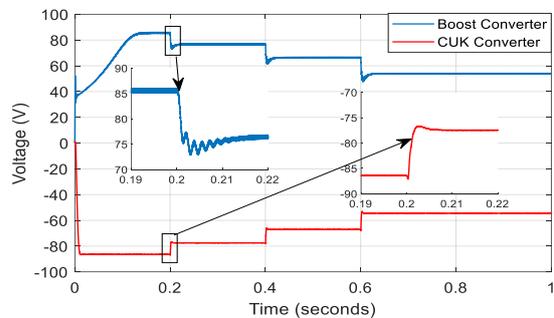


Figure 12. Boost converter and CUK converter output voltages

The variation of the load voltage obtained depending on the irradiance value with respect to time is given in Table 5.

Table 5. The performance of Boost and CUK converter topologies

Time	Irradiance Value	Boost Converter	CUK Converter
0-0.2s	1000 W/m^2	85.6V	-86.5V
0.2-0.4s	800 W/m^2	76.6V	-77.5V
0.4-0.6s	600 W/m^2	66.5V	-67V
0.6-1s	400 W/m^2	54V	-54.5V

In Figure 13, the variation of the power consumed on the load connected to the output in DC-DC converter topologies according to time is given.

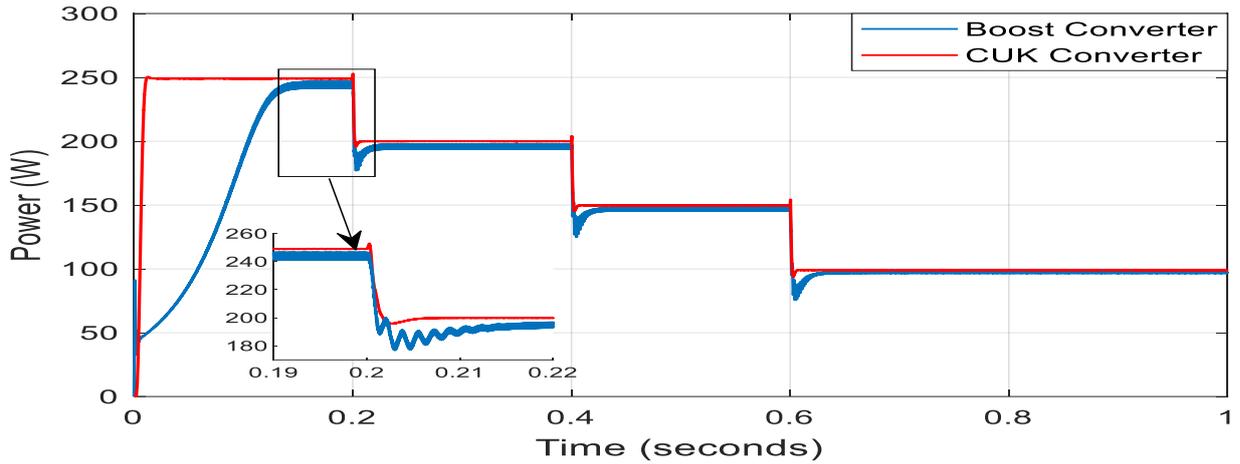


Figure 13. Boost converter and CUK converter output powers

In Table 6, the power values transferred to the load for the first scenario are shown in comparison.

Table 6. The performance of Boost and CUK converter topologies

Time	Irradiance Value	GMPP	Boost Converter	CUK Converter
0-0.2s	1000 W/m ²	249.83W	244.5W	249.12W
0.2-0.4s	800 W/m ²	201.56W	196 W	200.5W
0.4-0.6s	600 W/m ²	152.01W	147W	150.6W
0.6-1s	400 W/m ²	101.36W	98W	99.5W

When the results of both converters are examined, it is seen that the CUK converter circuit topology reaches the MPP point in a shorter time compared to the boost converter topology for the first scenario. In addition, it has been observed that there is less oscillation in the current, voltage and power values of the CUK converter circuit topology at the MPP point. The performance criteria such as settling time, rise time and oscillation (output power) of these two converters are shown in Table 7 comparatively.

Table 7. Comparison of Boost and CUK converter performance (0-0.2s)

Performance Metrics	Boost Converter	CUK Converter
Settling time (sec)	132 ms	10 ms
Rise time (sec)	110 ms	5.14 ms
Oscillation (%) (Power)	% 1.78	% 0.22

Second Scenario: In the second scenario, the effect of load change on converter performance (current, voltage and power) was examined by keeping the irradiance value and temperature constant. For this purpose, the irradiance value is determined as 700 W/m² and the temperature was determined as 25°C fixed values. The situation where the load value changes at intervals of 0.3s has been examined. The load value was determined as 20Ω in the 0-0.3s time interval, the load value in the 0.3-0.6s time interval as 25Ω and the load value in the 0.6s and later time period as 30Ω.

Figure 14 shows the variation of the currents of the boost converter and CUK converter circuit topologies for the load change situation at certain time intervals.

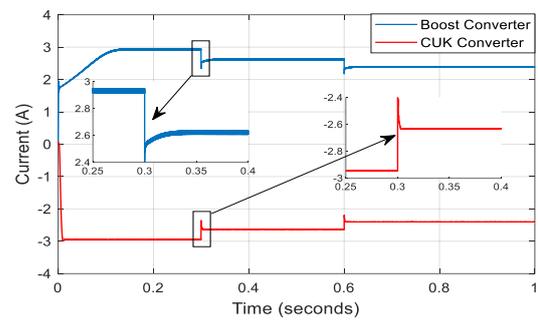


Figure 14. Boost converter and CUK converter output currents

As the load value connected to the output of the converter changes, the current value also changes accordingly. The current values obtained for this scenario are shown in Table 8.

Table 8. The performance of Boost and CUK converter topologies

Time	Irradiance Value	Load	Boost Converter	CUK Converter
0-0.3s	700 W/m ²	20Ω	2.98A	-2.95A
0.3-0.6s	700 W/m ²	25Ω	2.62A	-2.63A
0.6-1s	700 W/m ²	30Ω	2.39A	-2.4A

Figure 15 shows the time-dependent variation of the output voltages of the second scenario boost converter and CUK converter.

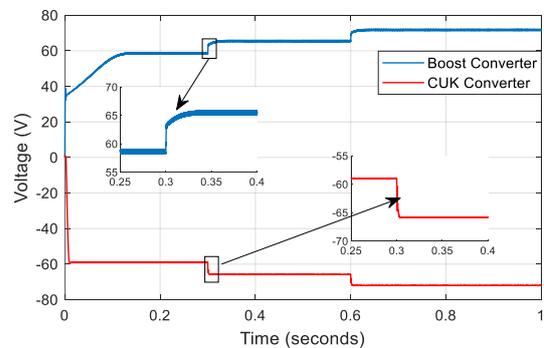


Figure 15. Boost converter and CUK converter output voltages

The change in the current value affects the voltage value so that the power transferred to the load remains constant. The output voltage values of both converters for 700 W/m² fixed irradiance value and variable load values are given in Table 9.

Table 9. The performance of Boost and CUK converter topologies

Time	Irradiance Value	Load	Boost Converter	CUK Converter
0-0.3s	700 W/m ²	20Ω	58.6V	-59V
0.3-0.6s	700 W/m ²	25Ω	65.5V	-65.8V
0.6-1s	700 W/m ²	30Ω	71.8V	-72.1V

Figure 16 is given for the type converter and the CUK converter topology that amplify the time-dependent variation of the power dissipated on the load.

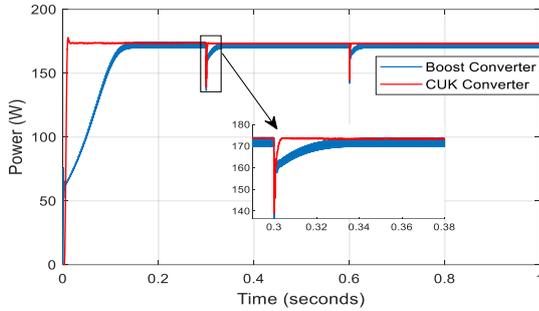


Figure 16. Boost converter and CUK converter output powers

In Table 10, the power values transferred to the load for the second scenario situation are presented.

Table 10. The performance of Boost and CUK converter topologies

Time	Irradiance Value	Load	GMPP	Boost Converter	CUK Converter
0-0.3s	700 W/m ²	20Ω	176.941 W	171.6W	174W
0.3-0.6s	700 W/m ²	25Ω	176.941 W	171.6W	174W
0.6-1s	700 W/m ²	30Ω	176.941 W	171.6W	174W

It has been observed that the CUK converter circuit topology reaches the MPP point quickly for the situation where the load value changes at certain intervals, while the boost converter topology cannot fully reach the MPP point and oscillates more in current, voltage and power values. DC-DC converters performance parameters rise time, settling time and oscillation (%) for the two designed models are given in Table 11.

Table 11. Comparison of Boost and CUK converter performance (0-0.2s)

Performance Metrics	Boost Converter	CUK Converter
Settling time (sec)	122 ms	9 ms
Rise time (sec)	90 ms	4.1 ms
Oscillation (%) (Power)	% 2.298	% 0.115

Third Scenario: As the last scenario, the difference in the irradiance value coming to the panel depending on the sun's condition during the day has been examined. In order to evaluate this situation, a function whose irradiance value changes trapezoidal is chosen. The variation of irradiance with time for this scenario is shown in Figure 10b. At 700 W/m², 25 °C, GMPP is 50.027W, while for 1000 W/m², 25 °C, the GMPP is 249.83W. The boost converter output power is given in Figure 17.

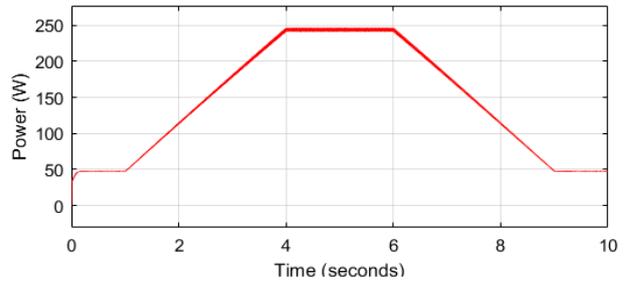


Figure 17. Boost converter output power

The irradiance value was determined as 200 W/m² in the 0-1s time interval. In this case, the settling time is 0.123s and the output power is 47.5W. In the 1-4s time interval, the irradiance value increased linearly. Depending on the increase in irradiance, the power value transferred to the load also increased. In the case of 1000 W/m² irradiance in the 4-6s time interval, the power value is 244.5W. While the irradiance decreases linearly between 6-9s, the power value also decreases as the power value changes directly proportional to the irradiance. Finally, when the irradiance value is considered 200 W/m² again, the output power value is measured as 47.5W.

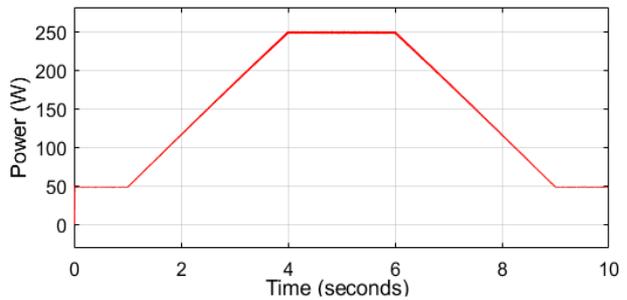


Figure 18. Boost converter output power

The CUK converter output power is seen in Figure 18. The output power value transferred to the load in the 0-1s time interval was measured as 48W in 0.016s. Figure 18 shows an increasing power change in response to the increasing irradiance value in the 1-4s time interval, and a decreasing power change depending on the decreasing irradiance value in the between 6-9s. The power is 249.6W in the between 4-6s. Finally, when the irradiance value is considered as 200 W/m² again, the power value is measured as 48W.

4. CONCLUSION

Solar energy systems are the most preferred renewable energy source due to their low maintenance costs. In order to adjust the output voltage of the PV panel to the desired values, booster type and CUK converter topologies are used. In this study, the PSO-based NN optimization algorithm was applied to both converter topologies. The performance of the converters for different scenarios was compared according to parameters such as settling time, rise time and oscillation. Firstly, DC-DC converter topologies for variable irradiance conditions are compared and performance criteria are presented. As the second scenario, the performance evaluation of the converter topologies according to the current, voltage and power parameters is made for the variable load situation.

In this situation, the CUK converter topology has reached its MPP by responding faster to all load change situations. Since the angle of incidence of the sun rays cannot reach the panels at the same rate during the day, the irradiance value is determined as a trapezoidal function in order to evaluate this scenario. The performance of the DC-DC converters under these scenarios is investigated. It is observed that the CUK converter responds faster to changes and reaches MPP earlier than the boost converter. In studies conducted for three different cases, it is observed that the CUK converter reaches the MPP faster and oscillates less than the boost converter. Matlab/Simulink program is used for simulation studies. PSO algorithm is frequently preferred in optimization problems due to its advantages such as simple structure and easy applicability. In this study, the panel data is trained with a neural network based on the PSO optimization algorithm. CUK converter circuit topology has been applied to this method, which is mostly used with boost converter structure in the literature, and it has been tried to contribute to the relevant field and successful results have been obtained. As a future research, the application of different optimization algorithms that have not yet been applied in this field to these systems will be studied.

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